# Integrated Current Sensor FHS AH 600

## **Definition**

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FHS AH is a coreless current sensor. It is an open loop field sensor, surface mounted device based on the Hall effect principle. The device uses an Integrated Magnetic Concentrator combined with improved sensitivity, offset and temperature stress compensation allowing high accuracy and low noise current sensing. FHS AH sensor is highly suitable for DC, AC, pulsed and mixed current measurements with 160 kHz bandwidth and fast response time  $\leq 2 \ \mu s$ .

The contactless current sensing sollution allows no additional power losses and high dielectric insulation between the primary circuit and transducer electronics.

FHS AH removes the need of an exernal magnetic core, offers flexibility in compact design and it is a perfect solution for the current measurements through a conductor as PCB trace.

## Main features & advantages

- Hall effect current sensor with Integrated Magnetic Concentrator (IMC)
- Magnetic field measurement range ±3.3 mT
- Supply voltage 5 V
- Low power consumption
- Operating temperature range: -40 ... 125 °C
- High bandwidth: 160 kHz
- Excellent linearity < 0.5 % Full Scale
- Sensitivity error < 1 % from −40 ... 125 °C
- Insulated current measurement
- Small footprint with standard SOIC8 surface mount PCB
- Low cost, Small size
- Internal reference voltage.

## **Typical applications**

- Small drives
- BMS
- Motor control
- UPS
- HVAC
- White goods
- PCB track current sensing.





Figure 1: FHS AH package - SOIC-8

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## Application circuit and pinout



Figure 2: Application circuit

## **Table 1: Pin definition**

Pin#	Name	Description					
1	$U_{\mathrm{ref}}$	eference voltage					
2	$U_{\rm out}$	itput voltage					
3	GND	round terminal					
4	U <sub>c</sub>	Supply voltage					
5	NC	Not connected or GND					
6	OCD <sub>E</sub>	External OCD					
7	OCD <sub>1</sub>	Internal OCD					
8	$U_{\rm E \; OCD}$	External OCD threshold voltage terminal					

## **Polarity**

 $U_{\rm out}$  –  $U_{\rm ref}$  is positive following the direction of the arrow.



## **Table 2: External circuit components**

External circuit example#	Unit	Value	Comment
$C_{ m supply}$	nF	47	< 100
C <sub>out</sub>	nF	4.7	< 6
C <sub>ref</sub>	nF	47	
R <sub>IOCD</sub>	kΩ	4.7	< 50
R <sub>E OCD</sub>	kΩ	4.7	< 50
R <sub>E</sub>	kΩ	100	
	kΩ	100	

 $C_{
m supply}$ ,  $C_{
m out}$  and  $C_{
m ref}$  should be mounted as close as to the pins. Ideally,  $R_{
m E\,ref}$  +  $R_{
m E}$  should have a value around 200 k $\Omega$  due to current limitation on  $U_{
m ref}$ .

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## Absolute maximum ratings

Parameter	Symbol	Unit	Rating
Ambient operating temperature	T <sub>A</sub>	°C	-40 +125
Ambient storage temperature	T <sub>Ast</sub>	°C	-55 +165
Maximum junction temperature	$T_{ m Jmax}$	°C	165
Output sink current		mA	-50
Output source current		mA	25
Magnetic flux density	B <sub>max</sub>	Т	±3

Stresses above these ratings may cause permanent damage.

Exposure to absolute maximum ratings for extented periods may degrade reliability.

## **Environmental and mechanical characteristics**

Parameter	Symbol	Unit	Value
Maximum supply voltage	$U_{\rm C\;max}$	V	8
Electrostatic discharge voltage (HBM-Human Body Model)	$U_{\rm ESD\;HBM}$	kV	2
Electrostatic discharge voltage (CDM -Charged Device Model)	$U_{\rm ESD\;CDM}$	V	500
Mass	т	g	0.08

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## **Electrical data**

At  $T_{\rm A}$  = 35 °C,  $U_{\rm C}$  = +5 V,  $R_{\rm L}$  = 100 k $\Omega$  unless otherwise noted.

Parameter	Symbol	Unit	Min	Тур	Max	Conditions	
DC supply voltage	Uc	V	4.5	5	5.5	−40 °C 125 C	
Current consumption	I <sub>c</sub>	mA	12	14	19	-40 °C 125 C; without R <sub>L</sub>	
Magnetic flux density measuring range	B <sub>M</sub>	mT		±3.3		U <sub>c</sub> > 4.6 V	
Output voltage in a flux density <i>B</i>	$U_{\rm out}$	V	0.5		4.5	$U_{ref} + U_{OE} + (S_N \times B)$ -40 °C 125 C	
Linearity error 0 B <sub>M</sub>	ε <sub>L</sub>	% of $B_{\rm M}$	-0.5		0.5		
Nominal sensitivity	S <sub>N</sub>	mV/mT		600			
Sensitivity error	$\varepsilon_s$	%	-1	±0.5	1		
Temperature coefficient of S	TCS	ppm/K	-200		200	−40 °C 125 °C, referred to 35 °C	
Lifetime sensitivity drift	SL	%		±1.5			
Internal reference voltage @ $B = 0 T$	$U_{\rm Iref}$	V	2.48	2.5	2.52		
Output internal resistance of $U_{\rm ref}$	R <sub>I ref</sub>	Ω	120	200	333	−40 ° C +125 °C	
Electrical offset voltage	U <sub>oe</sub>	mV	-10		10	$U_{\rm out}$ – $U_{\rm ref}$	
Lifetime offset drift	U <sub>oel</sub>	mV		±15		$U_{\rm out}$ – $U_{\rm ref}$	
Remanent field/Magnetic offset	U <sub>om</sub>	mV	-9		9	After $\pm B_{M}$ , on FS	
Temperature coefficient of $U_{\rm ref}$	TCU <sub>ref</sub>	ppm/K	-150		150	−40 ° C +125 °C; referred to 35 °C	
Temperature coefficient of $U_{0E}$	TCU <sub>OE</sub>	mV/K	-0.1		0.1	-40 ° C +125 °C; referred to 35 °C	
Output internal resistance of $U_{\rm out}$	R <sub>out</sub>	Ω			5	DC	
Noise voltage spectral density	U <sub>no</sub>	µV/Hz 1/2		17			
Delay time to 10 % of the final output value for $B_{\rm N}$ step	t <sub>D 10</sub>	μs			2	@ input signal 50 A/2 μs	
Delay time to 90 % of the final output value for $B_{\rm N}$ step	t <sub>D 90</sub>	μs			2	@ input signal 50 A/2 μs	
Eroquonov hondwidth	DW	kHz		160		@ -3 dB	
	DVV			100		@ -1 dB	

### Definition of typical, minimum and maximum values

Unless otherwise stated (e.g. "100 % tested"), the LEM definition "min" and "max" is that the probability for values of samples to lie in this interval is 99.73 %. For a normal (Gaussian) distribution, this corresponds to an interval between -3 sigma and +3 sigma. The definition of "typical" is that the probability for values of samples to lie in this interval is 68.27 %, corresponding to an interval between -sigma and +sigma for a normal distribution.

Typical, maximal and minimal values are determined during the initial characterization of the product.

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### **Overcurrent detection (OCD)**

Overcurrent detection is a feature included on FHS AH product in order to detect high peaks of currents happening during operation.

Two overcurrent detection types are included in this product: Internal OCD and External OCD.

At  $T_{\rm A}$  = 35 °C,  $U_{\rm C}$  = +5 V,  $R_{\rm L}$  = 100 k $\Omega$  unless otherwise noted.

Parameter	Symbol	Unit	Min	Тур	Мах	Comment
	IN	TERNAL	OCD			
Internal OCD detection threshold	B <sub>I OCD Th</sub>	mT		4.8		1.45 × B <sub>M</sub> , -40 ° C +125 °C
Internal OCD threshold error	€ <sub>I OCD Th</sub>	%	-10		10	Referred to $B$ @ $U_{out} - U_{ref} = 0.8 V$
Internal OCD output on resistance	$R_{ m on \ I \ OCD}$	Ω	70	95	150	Open drain output, active low
Internal OCD output hold time	t <sub>hold I OCD</sub>	μs	7	10	14	
Internal OCD delay time	t <sub>DIOCD</sub>	μs	1.3		2.1	
	EX	TERNAL	OCD			
External OCD detection threshold	B <sub>E OCD Th</sub>	т				Calculated from external resistors - see formula below
External OCD threshold error	$\varepsilon_{E \ OCD \ Th}$	%	-5		5	Referred to $B @ U_{out} - U_{ref} = 0.8 V$
External OCD output on resistance	R <sub>on E OCD</sub>	Ω	35	200	280	Open drain output, active low
External OCD output hold time	t <sub>hold E OCD</sub>	μs		10		
External OCD delay time	t <sub>DEOCD</sub>	μs		10		

### Setting External OCD threshold:

The external Over-Current-Detection threshold is directly set by user with the external resistors  $R_{E}$  and  $R_{E ref}$  using the following formula:

$$U_{\rm E \ OCD} = \frac{R_{\rm E}}{R_{\rm F} + R_{\rm E \ ref}} \cdot U_{\rm ref}$$

The corresponding sensed magnetic field threshold is defined by the external voltage threshold and device sensitivity using the following formula:

$$B_{\text{E OCD}} = \frac{U_{\text{ref}} - U_{\text{E OCD}}}{S_{\text{N}}}$$
 with  $0.5 < U_{\text{E OCD}} < U_{\text{ref}} - 0.5$ 

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### **Terms and definitions**

### Sensitivity and linearity

To measure sensitivity and linearity, the primary current (DC) is cycled from 0 to  $B_{\rm M}$ , then to  $-B_{\rm M}$  and back to 0 (equally spaced  $B_{\rm M}$  /10 steps). The sensitivity *S* is defined as the slope of the linear regression line for a cycle between  $\pm B_{\rm M}$ .

The linearity error  $\varepsilon_1$  is the maximum positive or negative difference between the measured points and the linear regression line,

expressed in % of  $B_{\rm M}$ .

## **Thermal coefficient of Sensitivity**

$$TCS(T) = \left(\frac{\varepsilon_{\text{S at }T} - \varepsilon_{\text{S at RT}}}{T - \text{RT}}\right) \times 10^4 \text{ ppm/K}$$

### Thermal coefficient of $U_{ref}$

$$TCU_{ref}(T) = \left(\frac{U_{ref at T} - U_{ref at RT}}{U_{ref at RT}}\right) \times \frac{1}{(T - RT)} \times 10^{6} \text{ ppm/K}$$

RT: Room Temperature *T*: temperature which *TCS* is calculated.

### **Electrical offset voltage**

The electrical offset voltage  $U_{OE}$  is the residual output voltage when the input signal is zero.

### Magnetic offset voltage

The magnetic offset voltage  $U_{OM}$  is the consequence of any magnetic flux density over the stated maximum measurement range.

#### **Delay times**

The delay time  $t_{D 10}$  @ 10 % and the delay time  $t_{D 90}$  @ 90 % with respect to the primary are shown in the next figure. They are measured at nominal current with a current rise time of 2 µs.



Figure 3: Delay time  $t_{D10}$  @ 10 % and delay time  $t_{D90}$  @ 90 %

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## **Internal OCD behavior**



Figure 4: Internal OCD behaviour

## External OCD behavior





Figure 5: External OCD behaviour

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## Accuracy



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## 

## **FHS AH 600**

## Noise



Figure 6: Noise voltage spectral density



di/dt







Figure 7: Sensitivity and phase over frequency







Figure 10: delay time to current step (d*i*/d*t*)

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## **Application information**

### Basic operation: example with a long thin conductor

FHS AH is a galvanically insulated current transducer.

It senses the magnetic field generated by the measured current and transforms it into an output voltage. If the current is bidirectional, FHS AH will sense the polarity of the magnetic field and generate a positive or negative output voltage relative to the reference voltage.

A simple case is presented which illustrates the current to magnetic field and then to output voltage conversion.

A current flowing in a long thin conductor generates a flux density around it:  $B(T) = \frac{\mu_0}{2\pi} \cdot \frac{I_P}{r}$  T

With:

 $I_{\rm p}$  the current to be measured (A) r the distance from the center of the wire (m)  $\mu$ 0 the permeability of vacuum (physical constant,  $\mu$ 0 = 4. $\pi$ . 10-7 (H/m)





If FHS AH is now placed in the vicinity of the conductor (with its sensitivity direction colinear to the flux density B), it will sense the flux density and the output voltage will be:

$$U_{\text{out}} = S_{\text{N}} \cdot B = S_{\text{N}} \cdot \frac{\mu_{0}}{2\pi} \cdot \frac{I_{\text{P}}}{r} = 1.2 \cdot 10^{-4} \cdot \frac{I_{\text{P}}}{r} \text{ V}$$

where  $S_{N}$  is the FHS AH sensitivity (600 V/T).

The transducer sensitivity to current is therefore:

$$S = \frac{U_{out}}{I_P} = \frac{1.2 \cdot 10^{-4}}{r} = (V/A)$$

The next graph shows how the output voltage decreases when r increases.

Note: that the sensitivity also depends on the primary conductor shape.





Figure 12: Sensitivity versus the distance between the conductor and the FHS AH sensing elements

The example above is of limited practical use as most conductors are not round and thin but explains the principles of FHS AH operation.

The measuring range limit ( $I_{\rm PM}$ ) is reached when the output voltage ( $U_{\rm out}$  –  $U_{\rm ref}$ ) reaches 2 V.

This limit is due to electrical saturation of the output amplifier.

The input current or field may be increased above this limit without risk for the circuit.

Recovery will occur without additional delay (same delay time as usual).

The maximum current that can be continuously applied to the sensor  $(I_{PM})$  is only limited by the primary conductor carrying capacity.

#### Single track on PCB

The main practical configuration for FHS AH is to measure current flowing directly through a PCB track (single or multi-layers), offering the following advantages:

- Insulation is guaranteed by PCB design. Very high insulation can be achieved by placing the primary (current carrying) track on the opposite side of the PCB.
- Cost effective solution
- Wide input current range (up to 100 A)
- Compact solution
- No added power losses
- No change on current track impedance



Figure 13: Principle of FHS AH used to measure current in a PCB track





Figure 14: Sensitivity versus track width and versus distance between the track and the FHS AH sensing elements (TBC).

The sensitivity depends on the track width and distance, as shown in above figure.

The maximum current that can be safely applied continuously is determined by the temperature rise of the track. The use of a track with varying width gives the best combination of sensitivity and track temperature rise.

#### Multi-turns

For low currents (under 10 A), it is advisable to make several turns with the primary track to increase the magnetic field generated by the primary current.

#### **Busbar**

For very large currents (> 50 A), FHS AH can be used to measure the current flowing in a busbar. The position of FHS AH relatively to the conductor has to be stable to avoid sensitivity variations.



Figure 15: Example with busbar

#### Accuracy considerations

Several factors influence the output accuracy of FHS AH as a current sensor:

- The sensitivity of the FHS AH
- · The distance and shape of the primary conductor
- The circuit output offset
- The circuit non-linearity
- Stray fields.

The sensitivity of the FHS AH is calibrated during production at 600 V/T.

As mentioned above, the distance and shape of the primary conductor also influence the sensitivity.

For best performance, no relative mouvement of the primary conductor to FHS AH should be possible.

Mechanical tolerances on device position and primary conductor design can have an impact on performance and part-to-part deviation and should therefore be minimized.

The magnetic fields generated by neighboring conductors, the earth's magnetic field, magnets, etc. are also measured if they have a component in the direction to which FHS AH is sensitive.

As a rule, the stronger the field generated by the primary current, the smaller the influence of stray fields and offset. The primary conductor should therefore be designed to maximize the output voltage.

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## PCB footprint & dimensions (in mm)



Notes:

All dimensions are in millimeters (angles in degrees)

\* Dimensions do not include mold flash, protrusions or gate burrs (shall not exceed 0.15 per side)

\*\* Dimension does not include interleads flash or protrusion (shall not exceed 0.25 per side)

\*\*\* Dimension does not include dambar protrusion.

Allowable dambar protrusion shall be 0.08 mm total in excess of the dimension at maximum material condition. Dambar cannot be located on the lower radius of the foot.

#### Soldering:

Recommended reflow soldering profile as standard: IPC/JEDEC J-STD-020 revision C.

### Typical connection diagram and ground plane

Good EMC practice requires the use of ground planes on PCBs. In drives where high dv/dt transients are present, a ground plane between the primary conductor and FHS AH will reduce or avoid output perturbations due to capacitive currents. However, the ground plane has to be designed to limit eddy currents that would otherwise slow down the delay time.

The effect of eddy currents is made negligible by cutting the copper plane under the package as shown in bellow figure.



Figure 16: Top side copper plane has a cut under the IC to optimize delay time

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### Tape & Reel

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<u>Plastic 13'' Reel</u>: Made by antistatic high-impact molded polystyrene. The mechanical integrity of the reel is not affected by humidity.



#### Moisture Barrier Bag (MBB):

Made in 5 different layers with total thickness of 0.18 mm. At the core is a layer of polyester sandwiched between aluminum shields. The outside layer: dissipative polyester, innermost layer: static dissipative polyethylene.



## Leader and Trailer



## **Carrier Tape Data**



## **Component Rotation and Lateral Movement**



Sketch A (Side or Front Sectional View) Component Rotation



Component Rotation



Sketch C (Top View) Component lateral movement

## **Plastic Reel Data**



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### Safety



If the device is used in a way that is not specified by the manufacturer, the protection provided by the device may be compromised. Always inspect the electronics unit and connecting cable before using this product and do not use it if damaged. Mounting assembly shall guarantee the maximum primary conductor temperature, fulfill clearance and creepage distance, minimize electric and magnetic coupling, and unless otherwise specified can be mounted in any orientation.



Caution, risk of electrical shock

This transducer must be used in limited-energy secondary circuits SELV according to IEC 61010-1, in electric/electronic equipment with respect to applicable standards and safety requirements in accordance with the manufacturer's operating specifications. Use caution during installation and use of this product; certain parts of the module can carry hazardous voltages and high currents (e.g., power supply, primary conductor).

Ignoring this warning can lead to injury and or/or cause serious damage.

De-energize all circuits and hazardous live parts before installing the product.

All installations, maintenance, servicing operations and use must be carried out by trained and qualified personnel practicing applicable safety precautions.

This transducer is a build-in device, whose hazardous live parts must be inaccessible after installation.

This transducer must be mounted in a suitable end-enclosure.

Besides make sure to have minimum 30 mm between the primary terminals of the transducer and other neighboring components. Main supply must be able to be disconnected.

Always inspect the flexible probe for damage before using this product.

Never connect or disconnect the external power supply while the primary circuit is connected to live parts.

Never connect the output to any equipment with a common mode voltage to earth greater than 30 V.

Always wear protective clothing and gloves if hazardous live parts are present in the installation where the measurement is carried out.

This transducer is a built-in device, not intended to be cleaned with any product. Nevertheless, if the user must implement cleaning or washing process, validation of the cleaning program has to be done by himself. When defining soldering process, please use no cleaning process only.



ESD susceptibility

The product is susceptible to be damaged from an ESD event and the personnel should be grounded when handling it.

Do not dispose of this product as unsorted municipal waste. Contact a qualified recycler for disposal.

Although LEM applies utmost care to facilitate compliance of end products with applicable regulations during LEM product design, use of this part may need additional measures on the application side for compliance with regulations regarding EMC and protection against electric shock.

Therefore, LEM cannot be held liable for any potential hazards, damages, injuries or loss of life resulting from the use of this product.

## **Version history**

Date	Version	Comments
8 June 2023	0	New datasheet
3 July 2024	1	Modified definition paragraph (page 1); Modified value for Cref symbol (page 2); Removed first sentence in Absolute maximum ratings paragraph and added $U_{\text{ESD CDM}}$ sentence (page 3); Added 2 paragraphes (elecrical offset voltage and Magnetic offset voltage) modified curve for Delay times paragraph (page 6); Removed first curve (page 8); Replaced some curves (page 9 and 10); Modified Version history (page 17).

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